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Estimate of the Longitudinal Coupling Impedance for the NSNS Accumulator Ring

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# BNL/NSNS TECHNICAL NOTE

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A. G. Ruggiero and M. Blaskiewicz

April 21, 1997

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## Estimate of the Longitudinal Coupling Impedance for the NSNS Accumulator Ring\*

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March 25, 1997

#### Introduction

The concept of longitudinal coupling impedance was introduced for the first time by Sessler and Vaccaro in 1967 [1] when the wall electromagnetic properties were approximated by a sequence of lumped components. It was subsequently re-examined in 1970-71 [2] in terms of the surface characteristic impedance. A similar concept was also adopted by Sacherer [3] in 1974. In this report we shall follow the concept expressed in these references. The definition relies on the smooth, homogeneous, and continuous properties of the vacuum chamber wall, as in the case of the resistivity of the wall. It was pointed out [2,4] that a beam of charged particles generates two types of electromagnetic fields that can act back on the beam. The two contributions add to each other. The first represents the electromagnetic field stored in the space between the beam and the vacuum chamber wall. It is independent of the wall properties, and primarily depends on the transverse dimensions of the beam and pipe. This contribution has an electric and a magnetic part which have tendency to cancel with each other, by exhibiting a  $\gamma^2$  - dependence, where  $\gamma$  is the ratio of the total energy to the rest energy of a particle in the beam. This contribution is referred to as the "space-charge". The second contribution is proportional to the velocity Bc of the beam, and it is otherwise independent of the beam energy. Nevertheless, this term depends very intimately on the electric properties of the vacuum chamber wall, and of the components located in its proximity. It represents the effect of the induced beam current flowing longitudinally along the wall and generating longitudinal fields as it encounters pipe resistivity and discontinuities.

The analytical estimate of the longitudinal coupling impedance relies on some model of the surrounding and on some approximation, for instance those relevant to the transverse geometry of the vacuum chamber and of the beam distribution. We shall follow below the circular geometry approximation [2,4], though other geometries have also been considered in the past: rectangular [4], elliptical [5], and axially displaced beam [6]. The theoretical model developed in [1-4] is essentially correct in the case of smooth vacuum chamber walls made of homogeneous and isotropic material. Moreover, it strictly applies only to the case the surface characteristic impedance is uniform and continuous along the wall, as in the case of the wall resistivity. The model also deals with a straight cylindrical pipe, whereas in reality the beam and the vacuum chamber are bent to form a toroidal shape. This may create a resonating behavior between the beam and the pipe which may be very important in high-energy storage rings [7]. Fortunately, this effect can be entirely ignored in the case of a relatively small storage ring and a sufficiently low-energy proton beam.

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All the other vacuum chamber components are really discontinuities of the vacuum pipe. They do not have then the smooth behavior of the resistivity of the wall. One can calculate wake fields left behind by the beam and average them out over the entire circumference. This approach, which is quite common and exploited, has been question in the past, for instance, in the case of striplines [8]. It seems that the actual distribution of identical devices around the ring may affect the frequency contain of the response triggered by the beam itself, which is the common source of fields to all the devices. Discontinuities of the vacuum chamber that are usually analytically estimated are: striplines for clearing electrodes and beam position monitors, vacuum chamber steps, bellows, vacuum ports, and RF cavities. A method was developed [7,9,10] to determine the longitudinal coupling impedance limit in a given particle accelerator, by gathering all the analytical equations available in the literature which provide the frequency dependence of the coupling impedance from different vacuum chamber components. This method is still in use these days. Alternatives are the use of numerical computer programs, like MAFIA [11] and ABCI [12], for more detailed evaluation, or measurement of the various vacuum components on a bench using the wire method. The beam itself, when it is available, may give information of the coupling to the surrounding by studying its behavior under controlled excitation.

A very important consideration is the nature of the vacuum chamber cutoff. Above a frequency which corresponds to a wavelength comparable or smaller than the vacuum chamber size, the longitudinal coupling impedance effectively vanishes [2,13], the beam is electromagnetically coupled to the free space, and the details of the vacuum chamber are no longer relevant. Even the free-space coupling impedance nevertheless has a cutoff-of its own. In the case of low-energy proton accelerators, the presence of the vacuum chamber wall completely screens the beam from the free-space and the beam is inhibited from radiating.

This technical note reports on the results of investigation of the longitudinal coupling impedance in the NSNS Accumulator Ring. Conventional formulae that can be found in the literature have been used, some of them with arguable validity. It is customary to define the ratio Z/n as the longitudinal coupling impedance where n is the harmonic number, the ratio of the frequency of interest to the revolution frequency.

#### The NSNS Accumulator Ring

The function of the Accumulator Ring is to take the 1.0 GeV proton beam from the Linac and convert the long Linac beam pulse of about 1 ms into a 0.5 second beam in about 1280 turns. The bunch compression occurs during the injection process, and the beam is immediately extracted at the end of the process. The final beam has an intensity of  $2.08 \times 10^{14}$  proton per pulse, resulting in 2 MW average beam power at 60 Hz repetition rate.

The lattice of the Accumulator Ring is a simple FODO lattice with four-fold symmetry [14], and the dispersion function is reduced to zero at straight sections by the missing magnet scheme. The total circumference of the ring is 220.7 m and the transition energy is  $\gamma_T = 4.93$ , higher than the operating energy of 1 GeV. The salient design parameters are shown in Table 1

Average Power	2 MW
Kinetic Energy	1.0 GeV
Circumference, $2\pi R$	220.7 m
Bending Field	7.4 kG
Number of Protons, N	$2.08 \times 10^{14}$
Betatron Tunes, Q <sub>H/V</sub>	5.8 / 5.8
Transition Energy, $\gamma_{T}$	4.93
Natural Chromaticity, $\xi_{H,V}$	-6.50, -7.29
Full Betatron Emittance, $\varepsilon_{tot}$	$120 \pi \text{ mm mrad}$
Space-Charge Tune-Shift	< 0.2
RF peak Voltage (h=1)	42 kV
Revolution Frequency, f <sub>0</sub>	1.1887 MHz
Filling Time	1.018 ms
Synchrotron Period, T <sub>s</sub>	0.9 ms
Bunching Factor, B	0.325
Total Bunch Area, S	10 eV-s
Full Bunch Length, L	546.6 ns
Full Momentum Spread, $\Delta$	1.6 %
Average Beam radius, a	3.80 cm
Average Pipe Radius, b	10 cm

Table 1: General Parameters of the NSNS Accumulator Ring

#### The Ring Vacuum Pipe

The main parameters that enter the investigation are the dimension and the shape of the vacuum chamber, and the resistivity of the wall. The NSNS Accumulator Ring is made of three different sections: (i) The region of bending magnets where the shape of the vacuum pipe is rectangular with internal dimensions of 23 cm (H) and 13 cm (V). This region is expected to cover about 25 % of the whole ring circumference. (ii) The region of small quadrupoles where the vacuum pipe is circular with the internal diameter of 2b = 18 cm. This region is expected to cover about 60 % of the ring circumference. (iii) The region of large quadrupoles where also the pipe is circular but the internal diameter is 2b = 14 cm. This region covers the remaining 15 % of the circumference.

The analysis that follows applies strictly to circular geometry. Thus, we have approximated the rectangular vacuum chamber by a circular one with an internal diameter given by 2b = (H + V)/2 = 18 cm. By taking an average of the pipe dimension around the ring we have then assumed a pipe with an internal radius b = 10 cm. For most of the chamber components entering the analysis, it is indeed sufficient to specify a single shape and a single dimension.

The material of the vacuum chamber can be either stainless steel with a surface resistivity of  $\rho_w = 73 \,\mu\Omega x$  cm, or aluminum with  $\rho_w = 2.83 \,\mu\Omega x$  cm. The vacuum system may require in some places stainless steel for more rigidity of the vacuum chamber and to avoid electron desorption at the wall. On the other end, the resistive wall instability may be softened by employing a higher conductor like aluminum.

#### **Transverse Beam Dimension**

The betatron emittance quoted in Table 1 defines the total beam, that is 100% of it. It has the same value in the two planes of oscillation. We adopted the criterion to define the total emittance  $\varepsilon_{tot}$  as 5 times the rms emittance  $\varepsilon_{rms}$ . We also define the average values of the envelope functions  $\langle \beta_{H,V} \rangle = R / Q_{H,V}$ , with R the average closed orbit radius and  $Q_{H,V}$  the betatron tunes. The average rms beam size in the vertical and horizontal plane are then given by  $\sigma_{H,V} = (\langle \beta_{H,V} \rangle \varepsilon_{rms})^{1/2}$ . This is the contribution from the betatron motion alone, to which we should add the contribution in the horizontal plane from the relative momentum spread  $\delta$  in the beam, which is  $\sigma_E =$  $\langle \eta \rangle \delta$ , where  $\langle \eta \rangle = R / \gamma_T^2$  is the average value of the dispersion around the ring. The total beam relative momentum spread  $\Delta$  is given in Table 1. The rms value  $\delta$  is 1/5 of the full value.

It is sometime required to specify the average beam radius around the ring. The assumption commonly made is that the beam has a transverse uniform charge distribution and a circular shape with radius a, which we estimate to be  $a = 3 [\sigma_V (\sigma_H^2 + \sigma_E^2)^{1/2}]^{1/2}$ .

#### Low-Energy Proton Storage Rings

The frequency range and the magnitude of the wall-coupling impedance in a storage ring is determined essentially by the dimensions of the vacuum chamber and by the energy of the beam through the relativistic factor  $\gamma$ , the ratio of the total beam energy to the rest energy. A major feature of a low-energy storage ring is the low value of  $\gamma$  and therefore of the impedance frequency range of interest. In fact the cut-off harmonic number above which the beam does not interact effectively with the wall components [2] is given by  $n_c \sim \gamma R/b$ , where R is the average ring radius and b is the average vacuum chamber size. For a 1 GeV proton energy, a circumference  $2\pi R = 220.7$  m and a vacuum chamber radius b = 10 cm, we derive  $n_c \sim 726$ , which is a very narrow frequency range (of only 0.61 GHz) when compared to that of high-energy storage rings (SSC, LHC, RHIC, Tevatron, ...).

Next we quote contributions to the longitudinal coupling impedance in the frequency range below the pipe cut-off. The actual numerical estimate over the entire frequency range will then include proper weighting due to the roll-off functions of the cut-off which are calculated as shown in Table 2.

#### **Free-Space Impedance**

Another major feature when compared to electron beams at ultra-relativistic velocities [9,10], is the complete screening of the beam from interacting with the free space and therefore the inhibition of radiation. In fact the cut-off for synchrotron radiation is  $n_{rad} \sim 1.5 \gamma^3$ , considerably lower than the vacuum chamber cut-off  $n_c$ . In absence of the screening effect of a vacuum chamber, the free-space contribution to the longitudinal coupling impedance would otherwise be

$$Z/n = (177 \text{ ohm})(1.73 - i) n^{-2/3}$$
 (1)

Table 3 gives the estimate of the residual free-space impedance for the NSNS Accumulator Ring which includes the cutoff screening functions shown already in Table 2.

#### The Curvature of the Vacuum Chamber

A particle circulating in a toroidal vacuum chamber is subject to resonances. A wave propagating at the outer side of the beam, though speeding at higher velocity, has nevertheless a longer trajectory than the beam which is moving to the inside at slightly less velocity. Therefore it is possible for the beam to catch up with the wave it radiated the revolution (or few revolutions) before. This is a condition for the "pipe resonances" [7]. For a circular pipe it can be written as

$$\beta (1 + b/R) = 1 + 0.80862 \, n_r^{-2/3} \tag{2}$$

where  $n_r$  is the harmonic number value at resonance. It is immediately seen that one of the advantages of the low-energy storage ring is the absence of the vacuum chamber resonating modes that can be excited by the beam, because of the relative low value of the beam velocity  $\beta$ , and of the relatively large ratio of the pipe size b to the ring radius R.

#### **Space-Charge Contribution**

The space-charge contribution, that is the electromagnetic field stored in the region between the beam and the vacuum chamber, is [15]

$$Z/n = i Z_0 (1 + 2 \ln b/a) / 2 \beta \gamma^2$$
(3)

where  $Z_0 = 377$  ohm, and Z/n = 150i ohm, a purely "anti-inductive" contribution, which typically dominates over all other wall contributions in the case of a low-energy and relatively small proton storage ring.

The Space-Charge contribution to the NSNS Accumulator Ring is summarized in Table 4. Again, the roll-off due to the cutoff functions are also included.

Pipe Radius, b	10	cm
Beam Size, a	37.9615743	mm
a/b	0.37961574	
γ	2.06580266	
Radiation cutoff	13	10.9365323 MHz
Pipe cutoff	726	610.763268 MHz

n	Cut-Off	Cut-Off	Cut-Off	nb/γR		
	Free-Space	Space-Charge	Pipe			
1	0.99704579	0.99999986	0.99999905	0.00137741		
2	0.98823543	0.99999945	0.99999621	0.00275482		
3	0.97372416	0.99999877	0.99999146	0.00413223		
4	0.95376566	0.99999781	0.99998482	0.00550964		
5	0.92870467	0.99999658	0.99997628	0.00688705		
6	0.89896707	0.99999508	0.99996585	0.00826446		
7	0.86504789	0.9999933	0.99995352	0.00964187		
8	0.82749757	0.99999125	0.99993929	0.01101928		
9	0.78690719	0.99998893	0.99992316	0.01239669		
10	0.74389306	0.99998633	0.99990514	0.0137741		
20	0.30622598	0.99994532	0.99962062	0.02754821		
30	0.06975809	0.99987697	0.9991466	0.04132231		
40	0.00879363	0.9997813	0.99848334	0.05509642		
50	0.00061343	0.9996583	0.99763124	0.06887052		
60	2.368E-05	0.99950798	0.99659076	0.08264463		
70	5.0584E-07	0.99933037	0.9953625	0.09641873		
80	5.9796E-09	0.99912547	0.99394716	0.11019284		
90	3.9116E-11	0.9988933	0.99234554	0.12396694		
100	1.416E-13	0.99863388	0.99055855	0.13774105		
200	4.0198E-52	0.99454671	0.96276571	0.27548209		
300	2.288E-116	0.98777191	0.91816636	0.41322314		
400	2.611E-206	0.97836464	0.85917662	0.55096419		
500	0	0.9664011	0.78886707	0.68870523		
600	0	0.95197749	0.71069869	0.82644628		
700	0	0.93520869	0.62824278	0.96418733		
800	0	0.91622679	0.5449163	1.10192837		
900	0	0.89517929	0.46375906	1.23966942		
1000	0	0.87222728	0.3872713	1.37741047		

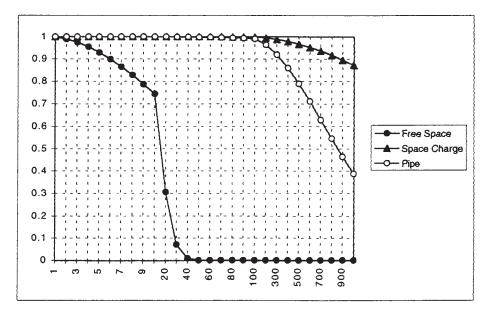


Table 3. Free-Space Residual Impedance (ohm)

10.9365323 MHz 610.763268 MHz		ohm
13 (n) 726 (n)	Imaginary	enon
t Full Screening	Real	726 none
cutoff F		" L
Radiation cutoff Pipe cutoff <i>F</i>		0
Radi Pipe		u/Z

Imaginary

Real

c

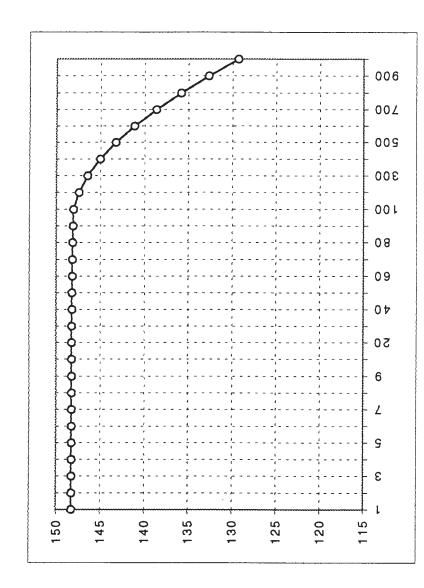
001 0 002 **4**00 001 ٥८ **0 †** 0 ٥١ L 7 0.005 0.004 0.003 0.002 0.001 -0.002 -0.003 -0.001 0

-0.0001674	.00041	-0.0007074	<b>T</b>	013	0164	19	8	-0.0024734	02690	02	0	00	-1.895E-05	ш	ц	-3.45E-10	-2.639E-12	-1.098E-14	-7.746E-53	-7.4E-117	-1.2E-206	0	0	0	0	0	0
0.00028996	.0007242		.00176	0.00230921	.002850	0.00336867	.003	S	.004660	0	0.00189032	0.00034958	3.2823E-05	Ч.	1.1	.97	90	σ.	41	Ņ	2.076E-206	0	0	0	0	0	0
-	2	ო	4	ۍ	9	7	80	თ	10	20	30	40	50		70	80	06				400		600	700	800	006	1000

148.266927 ohm

n Real Imaginary

148.266907	148.266846	148.266744	148.266603	148.26642	148.266197	148.265934	148.26563	148.265285	148.2649	148.25882	.24868	48.23		97	148.167643	.137	Ô	148.064377	147.458385	146.453905	145.059118	143.285321	141.146777	138.660519	135.84613	132.725483	129.322458
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
-	2	ო	4	S	9	7	8	6		20	30	40	_	_	70	80	06	100	200	ō	400	500	600	700	800	900	000



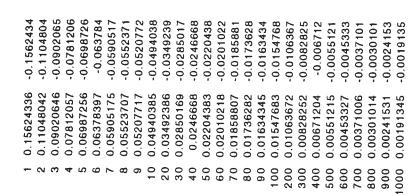
Z/n

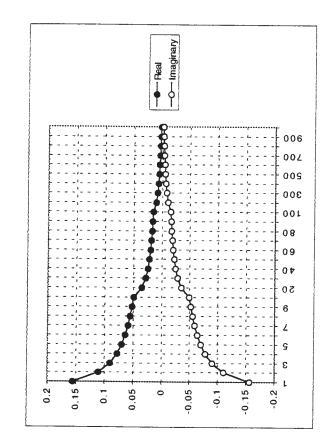
	Aluminum 2.83 μοhm-cm <b>0.85</b> 0.076532 mm @ n ≈ 1	75 μοhm-cm 5 mm 10 cm
	Copper 1.8 0.061036	3.687175 µohm-cm 5 mm 10 cm
220.688 m	Stainless Steel 7 3 0.15 0.38869706	lesistivity ber Thickness
Circumference	Material Resistivity Fraction Skin Depth	Average Wall Resistivity Vacuum Chamber Thickness Pipe Radius

Z/n @ n≈1 0.15624351 -0.1562435 ohm n<sup>-1/2</sup>

Real Imaginary

c





#### **Resistivity of the Vacuum Chamber**

The next contribution is the resistivity of the wall [4]

$$Z/n = (1 - i) (Z_0 \rho_w R / 2 b^2 n)^{-1/2}$$
(4)

The vacuum chamber is made 85% by Aluminum and 15% by Stainless Steel. For Aluminum  $\rho_w = 2.83 \mu$ ohm-cm and for Stainless Steel  $\rho_w = 73 \mu$ ohm-cm. The skin depth at the revolution frequency (n = 1) is 0.39 mm for Stainless Steel and only 0.08 mm for Aluminum. At the lowest harmonic n = 1, we have Z/n = (1 - i) 0.156 ohm as the total contribution of both types of vacuum chamber. The summary of the contribution from the resistivity of the wall is given in Table 5.

#### Vacuum Chamber Discontinuities

Next we have contributions which are caused by discontinuity of the vacuum chamber, like bellows, strip lines, vacuum chamber steps, vacuum pump ports, kicker magnets, and RF cavities. We shall examine them one-by-one below. The summaries of the details for each component are displayed in Tables 8 to 14.

#### Bellows:

Let M be the total number of bellows, m the number of convolutions per bellow, h the height and w the width of each convolution, then the contribution to Z/n in the low frequency range is [16]

$$Z/n = -i Z_0 (Mmw / 2\pi R) \ln (1 + h/b)$$
(5)

where  $Z_0 = 377$  ohm. The values of the parameters are as shown in the Table 6 below.

Total Number, M	48
Number of Convolutions / Bellow	6
Height of Convolutions, h	25 mm
Width of Convolutions, w	10 mm
Pipe Diameter	20 cm
Lowest Resonating Frequency	3.0 GHz
Z/n @ n = 1	- 1.1 i ohm

Table 6. Bellows in the NSNS Accumulator Ring

Moreover, each convolution resembles a cavity resonating at frequencies  $f_k = (1 + 2k) c/4h$  with

k = 0, 1, 2, ... Fortunately, with the parameters of the NSNS Accumulator Ring, the lowest resonating mode is well above the vacuum-chamber cut-off so that the beam is not expected to be capable to excite these modes. The bellows are made of stainless steel, have a circular geometry and are a natural extension of the vacuum chamber, without any re-entrance attached that may look like a resonating object. Screening of the bellows with metallic fingers does not seem to be required.

#### Strip Lines:

They can be beam position monitors, clearing electrodes, transverse damper devices, and ferrite-loaded kickers for extraction. The last component deserves an analysis apart, given the complexity of the electrical configuration.

We shall assume M strip lines each made of m plates of width w. The characteristic impedance is  $Z_{cha}$ . The downstream end of each plate is shorted, whereas the upstream end is terminated to the characteristic impedance. The general expression of the contribution to Z/n is [13]

$$Z/n = -4i Z_{cha} (Mm/n) (w/2\pi b)^2 \exp(-i \omega d/c) \sin (\omega d/c)$$
(6)

where d is the length of a plate, and  $\omega = n \beta c/R$ . Strip lines do resonate at the harmonic number  $n_{res} = c/(4d f_0)$ , which is at about the vacuum chamber cut-off. There are 48 beam position monitors located next to each quadrupole magnet. Each station is made of 4 plates, 20 cm long and 7.5 cm wide. The characteristic impedance  $Z_{cha} = 50$  ohm.

In the low frequency range Z/n = -i 0.7 ohm. The real part of the impedance peaks to about 0.5 ohm at the resonance. According to [8] the impedance given by Eq. (6) is to be multiplied by the following factor

$$\alpha_{\rm n} = (1 / {\rm M}) \sum_{\rm s} \exp(-i n z_{\rm s} / {\rm R})$$
 (7)

where  $z_s$  is the azimuthal location of the s-th plate around the circumference of the accelerator. For instance, in the case the position monitors have been distributed at equal distance, the factor  $\alpha_n$  is always zero, except for those harmonics n which are multiple of the number M. This effect, which we believe applies as well to all other short discontinuous components, can be taken as an advantage when deciding how to place them around the ring.

#### Vacuum Chamber Steps:

The longitudinal coupling impedance of M single steps (uncoupled) was estimated by H. Hereward [17]

$$Z/n = 2 M (1 - i \pi) Z_0 (W - 1)^2 b / 2 \pi^2 R$$
(8)

for  $n < n_W~=~2\pi R$  / 2b ( W - 1 ), and for  $n > n_W$ 

$$Z/n = Z_0 M (W-1) / 2\pi n$$
(9)

where  $W = b_2 / b_1$  is the ratio of the outer dimension  $b_2$  to the inner dimension  $b_1$  of the step.

According to this formula a substantial resistive contribution occurs in the low frequency range. It is associated to the actual energy loss suffered by a charged particle due to the diffraction phenomenon of an electro-magnetic plane wave being scattered by discontinuities. One should point out that very likely the result really applies only to wavelengths which have about the pipe dimension, that is in the proximity of the cut-off. Moreover, steps come in pair, an entrance followed by an exit discontinuity. Thus, one deals in reality with resonating cavities rather than single de-coupled steps. A step could be treated as a single discontinuity only at those wavelengths that are considerably shorter than the separation between the steps. The contribution represented by Eq.s (7 and 8) has been originally included in the analysis also for very low frequencies, but we believe that otherwise it should really not be included.

In any event, we have assumed 64 pairs of steps separated by about 2 m, with an inner radius of 10 cm and an outer radius of 13 cm. This yields  $n_w = 3678$  (4.4 GHz), which is well above the pipe main cut-off, and Z/n = (0.63 - 2.0 i) ohm.

When this geometry was investigated numerically with ABCI it was not possible to derive any significant result. Only when the outer dimension was increased to about 20 cm, the program gave significant results with resonances which nevertheless appeared only at very large frequencies, in proximity of 0.6 GHz and above. The results are shown in Figures 1 and 2. A modest inductive contribution Z/n = -i 0.15 ohm per cavity can be noticed in the low frequency range.

#### Vacuum Pump Ports:

These are circular openings of diameter d. The impedance is caused by the diffraction of the electro-magnetic wave through them. The impedance for M ports is [18]

$$Z/n = 2 M Z_0 \alpha^2 [n^3 + i(8/\pi)(n^2 n_{co} + n_{co}^3/3)]/3 \pi^3 R^4 b^2$$
(10)

where  $n_{co} = 2\pi R / d$  and  $\alpha = \pi d^3 / 16$ .

We count 48 vacuum port with an opening of d = 10 cm. The resistive contribution to Z/n of the ports is negligible in the low-frequency range, and increases to a maximum of about 1/3 ohm above the pipe cutoff. Whereas there is a substantial contribution reactive of about 10 ohm (positive).

#### Damper System:

This is made of two parallel striplines [19] 2.5 meter long and 10 cm wide, loaded with ceramic having a relative dielectric constant  $\varepsilon_r = 10$ . The characteristic impedance of each line is  $Z_{cha} = 50$  ohm. To allow for external powering, the plates are terminated at both ends to the characteristic impedance. The bandwidth is about 7 MHz centered to the resonating frequency of 10 MHz.

The impedance for such a device is [8,13]

$$Z/n = Z_{cha} (Mm/n) (w/2\pi b)^{2} [1 - exp(i \omega d/c) cos (\omega d/\beta c)]$$
(11)

where M = 1 and m = 2, and the other notations are as defined for the similar Eq. (6). The estimate of the impedance is shown in Table 12. Most of the contribution, as it is expected. is in the low frequency range around the resonating frequency.

#### Kicker Magnets:

There are 8 magnets each of about 40 cm length. They are made of two copper coils of narrow rectangular shape surrounded by ferrite. They can also be treated as transmission lines in the way similar to the one used for strip lines in general. The complication nevertheless arises from the uncertainty to determine how the currents induced by the beam on the coils flow in and out of the system. The kicker magnets are to be activated only for a very brief period of time at the end of the accumulation process just to quickly extract the beam, in one turn, from the Accumulator Ring. Thus, since our concern is the interaction with the beam during the accumulation time, we shall assume that the magnets can be designed with the provision to allow shorting of each copper plate to ground at the down stream end, a short that will be opened with a fast switch at the moment of firing the magnets. Moreover, we shall assume that the characteristic impedance of each plate to ground is 50 ohm and that the geometry and properties of the ferrite will match this value. The upstream end will then be matched to the characteristic impedance. With this arrangement one can estimate the coupling impedance by utilizing Eq. (6). The results are given in Table 13. Most of the contribution, as expected, is in the low frequency range, where nonetheless especially the resistive contribution is substantial.

Clearly the kicker magnets deserves a closer and more detailed examination in the future, especially when the properties of the ferrite and the characteristic impedance of the plates have been determined more accurately. Also it is important to devise an electric arrangement which will allow indeed shorting of the downstream ends during the accumulation process of the beam.

#### **<u>RF Cavity System</u>**:

An RF cavity can always be approximated with an equivalent RLC parallel circuit, and be described by the resonating frequency, the figure of merit Q and the shunt impedance  $R_s$ . The RF Compression system is made of two subsets: the primary one working at the revolution frequency, and the second one tuned to the second harmonic. Main parameters are given in the Table 7 below.

Table 7. The Compression RF System of the NSNS Accumulator Ring

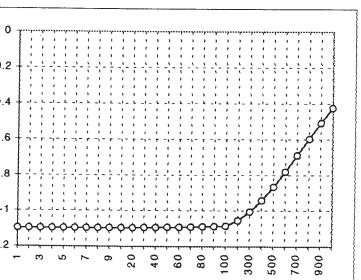
	System # 1	System # 2
Number of Gaps	6	2
Harmonic Number	1	2
Resonant Frequency	1.19 MHz	2.38 MHz
Q (unloaded)	200	200
Shunt Impedance	8 kohm	8 kohm

## Table 8: Bellows Coupling Impedance (ohm)

Circumference		220.688	3 m	
Number of Bellows No. of Convolutions Height of Convolution Width of Convolution Pipe inner Radius Pipe Cutoff		<b>10</b> 100		610.763268 MHz
Z/n @ n=1		-1.097841	ohm	
Capacitance / unit length Inductance / unit length Characteristic Impedance Resistivity Wall Impedance Q Shunt Impedance		2.78162897 707.499103 15.9482614 <b>73</b> 14.7988229 0.67712062 13.7495824	pH/cm ohm μohm-cm ohm/cm	stainless steel
Lowest resonating mode		2522	(n)	2997.925 MHz
Z/n @ resonance relative freq. spread Z/n with freq. spreading		none 10 none	ohm % ohm	
n Real		Imaginary		
1 2 3 4 5 6 7 8 9 10 20 30 40 50 60 70 80 90	000000000000000000000000000000000000000	-1.0978317 -1.0978244 -1.097815 -1.0978035 -1.09779 -1.0977744 -1.0977567 -1.0977369 -1.0974245 -1.0969041 -1.096176 -1.0952405 -1.0940982 -1.0927498 -1.091196	0 -0.2 -0.4 -0.6 -0.8 -1 -1.2	·
90 100 200 300 400 500 600 700 800 900 1000	000000000000000000000000000000000000000	-1.0894377 -1.0874758 -1.0569637 -1.0080007 -0.9432393 -0.8660506 -0.7802342 -0.6897107 -0.5982315 -0.5091337 -0.4251623		- w w w w w w w w w w w w w w w w w w w

900 1000

0 -0.4251623



Number No. of Plates Plate Length Plate Width Pipe Radius Characteristic	Impedance	48 4 20 cm 7.5 cm 10 cm 50 ohm	1
Z/n @ n=1		-0.6820926 ohm	
lowest resona	tina mode	315 (n)	374.740625 MHz
pipe cutoff	3	726 (n)	610.763268 MHz
n	Real	Imaginary	
1 2 3 4	0.01020323 0.01360343	-0.6820807 -0.6820448 -0.681985 -0.6819014	
5	0.01700288	-0.6817938	
6 7	0.02040138	-0.6816623	0.6
8	0.02379875 0.0271948	-0.681507	
9	0.03058935	-0.6813278 -0.6811247	0.4 + + + + + + + + + + + + + + + + + + +
10	0.0339822	-0.6808978	יוייייייייייייייייייייייייייייייייייי
20	0.06777624	-0.6773215	
30	0.10119542	-0.6713878	0.2
40	0.13405595	-0.6631366	
50	0.16617834	-0.6526234	0
60	0.1973888	-0.6399188	
70	0.22752055	-0.6251076	-0.2 + + + + + + +
80	0.2564151	-0.6082885	
90	0.2839234	-0.589573	-0.4 + + + + + + + + + + + + + + + + + + +
100	0.30990698	-0.5690846	4 4 4 4
200	0.46463296	-0.3000918	-0.6
300	0.41629577	-0.0311971	မဝင်ဝင်ဝင်ဝင်ဝင်
400	0.24410429	0.11014405	-0.8
500	0.07867929	0.10387108	
600	0 00359904	0.0000704	

6000.003599040.02387817000.01436041-0.03945498000.05237728-0.04622329000.06699292-0.015290710000.04908660.01381083

Table 9: Coupling Impedance from Beam Position Monitors (ohm)

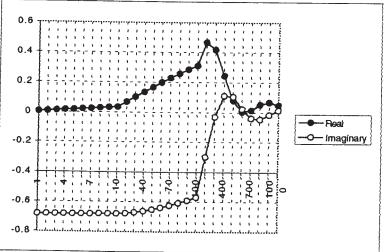


Table 10: Coupling Impedance of Vacuum Chamber Steps according to Hereward (ohm)

			4371.95251 MHz	5476 ohm	0 ohm	610.763268 MHz
64	10 cm	13 cm	3678 (n)	0.62641946 -1.96795476 ohm	0.00489408	726 (n)
Number	Pipe Radius inner	Pipe Radius outer	Cutoff mode	Z/n below cutoff	Z/n @ cutoff	pipe cutoff

Imaginary

Real

c

-1.96795289 -1.96794729 -1.96793796 -1.96792489 -1.96790809 -1.96788755 -1.96786329 -1.96783528 -1.96780355 -1.96776808 -1.96720816 -1.9662753 -1.96497005 -1.89467936 -1.23635336 -1.96329314 -1.96124552 -1.95604305 -1.95289114 -1.94937442 -1.80690986 -1.69082072 -1.39862286 -1.95882837 -1.55245471 -1.07237063 -0.91265684 -0.7621324 0.62641886 0.62641708 0.6264046 0.62640995 0.6261818 0.62546939 0.62493562 0.62641411 0.62639806 0.62639034 0.62638143 0.62637132 0.62636003 0.62428384 0.62351444 0.62262784 0.62162456 0.62050515 0.53820495 0.49416168 0.44519548 0.62588487 0.60309517 0.57515727 0.3935435 0.34134617 0.2905077 0.24259428 20 30 10 40 50 60 70 80 100 300 000 2 e 400000 06 200 400 500 600 700 800 006

Table 11: Coupling Impedance of Vacuum Ports according to Sands (ohm)

	2623 40024		57198 ohm
48	196.349541 cm <sup>3</sup>	1	9.8561E-10 8.99357198 ohm
10 cm	2207 (n)	10.5m	
Number	alpha	coherence	Z/n @ n≈1
diameter	cutoff mode	Pipe Radius	

Real Imaginary

c

8.99356253 8.99355078 8.9935312 8.9935312	.9934685 .9934254 .9933745 .9933745	8.99324928 8.99317488 8.99200009 8.99004176	.987299 .983772 .979459 .974359	8.961795 8.961795 9543269 .9543269 .8356618 .6351764	8.35024983 7.97972237 7.52562494 6.99451351 6.39807498 5.7528421 5.07905822
9.8561E-10 7.8849E-09 2.6611E-08 6.3078E-08	1.232E-0 .1288E-0 .3805E-0 5.046E-0	845 551 816 586	792. 0000 0000 0000	0.000975	0.30663055 0.14581298 0.20196822 0.25749809 0.30663055 0.34446129
<u>- 01 60 4</u>	5 8 7 6		40 50 70 80		500 500 600 800 900

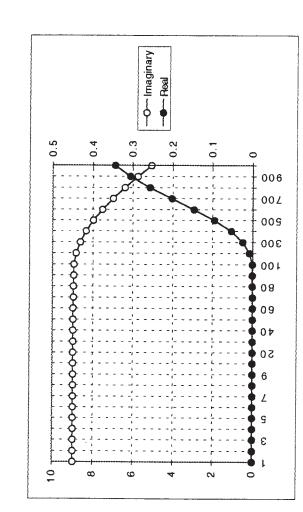


Table 12: Coupling Impedance (ohm) for the Damper System

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	271: 632:							1			1 ¦	2	5	- 1 					-	Z						
	9.48027125 MHz 610.763268 MHz								- / - /		· - 1		1  	~	ð				-	9						
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1 50 cm 10 cm 10 cm 10 cm 50 ohm	8 (n) (e (n)																									
1 250 cm 10 cm 10 cm 50 ohr	8 (n) 726 (n)	Ş	178	777	922	727	042	487	423	326	328	987	716	588	659	-12	771	905 818	311	-19	415	977	749	-18	058	501
		Imaginary	-0.48178	-0.4366777 -0.3667530	-0.2791922	0.1827727	-0.0868042 -1 7475-05	0.07043487	0.11974423	0.14588326	-0.0199328	0.02908987 0.05696027	0.00429716	0.0185588	0.03327659	-1.201E-17	0.00575771	0.959010.0	0.0012311	-5.874E-19	0.00175415	0.0026977	0.00160749	1.5312E-18	-0.00058	022
		<u>n</u>		ò ç	Ģ	Ģ	o i	0.07	0.11	0.14	98	0.0	0.0	0.0	0.03	7	0.0	- -	0	Ŷ	0.00	-0.0	0.00	1.5	Ÿ	0.00022501
			51	78	54	90	71	48	75	11	90	57	84	43	94	36	4 u V u	0 0	38.	04	90	37	87	05	54	97
90	e l	Real	0.11095551	0.21227678 0.29545	0.35405554	0.38446906	0.38620171 0.36184107	0.31660948	0.2576075	0.19285211	0.1066706	0.0632297	0.06091484	0.02351443	0.04980194	0.05107936	0.02554442	0.04100100 0.01010367	0.00652138	0.00424504	0.0057506	0.00300037	0.00066587	0.00328405	0.00072524	0.00098097
ns bedar	ыос	æ	.110	.212	.354	.384	386	316	0.25	.192	0.10	0.06 0.06	.060	.023	.049	.051	.025	019	900	.004	0.00	.003	000.	.003	000	000
yster lates 7 Imp	ating	£	1	0 N 0			0 0 9 1			0			0													
Number of Systems Number of Plates Plate Length Plate Width Pipe radius, b Permeability Dielectric Characteristic Impedance	lowest resonating mode pipe cutoff									-	20	30	50	60	20	80	60 1 1 0 0		300	400	500	600	700	800	006	1000

Table 13: Coupling Impedance (ohm) of the Kicker Magnets

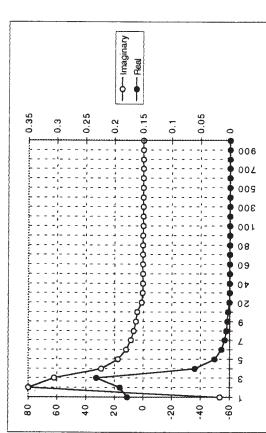
	5.92516953 MHz 610.763268 MHz		
4 2 40 cm 10 cm 10 cm 1 50 ohm	5 (n) 726 (n)	Imaginary	-2.977752 -2.4090452 -1.6060225 -0.7444274 -1.241E-16 0.49627553 0.68826923 0.68826923 0.68826923 0.68826923 0.68826923 0.68826923 0.68826923 0.68826923 0.68826923 0.68826923 1.2407E-16 1.23386-16 1.23386-16 1.23386-16 1.23386-16 1.23386-16 1.22316-16 1.22316-16 1.22316-16 1.22316-16 1.223576-16 1.223576-16 1.223576-16 1.223576-16 1.22666-16 1.22666-16 1.22666-16 1.22666-16 1.22666-16 1.22676-16 1.226666-16 1.226666-16 1.226666-16 1.226666-16 1.22666666666666666666666666666666666
Number of Systems Number of Plates Plate Length Plate Width Pipe radius, b Permeability Dielectric Characteristic Impedance	lowest resonating mode pipe cutoff	n Real	1         0.96753029           2         1.75027376           3         2.21050032           4         2.29111201           5         2.02637562           6         1.5737903           7         0.94732133           8         0.43754353           9         0.10749521           10         1.519385-32           9         0.10749521           10         1.519385-32           20         3.03785-32           9         0.10749521           10         1.519385-32           20         3.03785-32           9         0.10749521           10         1.505856-31           20         3.037856-32           30         4.55456-32           30         1.05876-31           90         1.208586-31           90         1.50516-31           90         1.50516-31           90         1.50516-31           90         1.50516-31           90         1.50516-31           90         1.50516-31           90         1.50516-31           90         5.22226-31           900 <td< th=""></td<>

Table 14: Coupling Impedance (ohm) of the RF Cavity System

System #2	~ ~	2.37735318 MHz 2.006+02	0.008 Mohm	0.00267785 mH	1673.65691 pF	-	7.4686753	0.56250625	0.177758 -53.332741
System #1	90 <del>~</del>	1.18867659 MHz 2.00E+02	0.008 Mohm	0.0053557 mH	3347.31381 pF	2	14.9373506	9.0001	0.2666637 79.9991111
	Number Harmonic Number	Resonant Frøquency Q	Shunt Impedance	Inductance	Capacitance	L	omega MHz	denominator	Z/n @ n ohm

n Real Imaginary

Ł 80 60 40 20 0 -20 -40 -60 -53.33269 79.9988076 61.9982133 29.3327115 11.8567153 8.55514662 6.47579103 5.07752812 0.08129059 0.04901803 0.03963106 0.00962819 17.6185757 4.09051861 1.00516254 0.44515447 0.15976164 0.11080003 0.06214306 0.00408084 0.00214797 0.0012622 0.00051285 0.00034057 0.24996474 0.00078967 0.00022902 0.00015491 1.136E-05 7.1418E-06 0.01357733 0.00813748 0.00528479 0.00363513 0.00261136 2.4363E-06 3.3E-08 8.0879E-09 4.5023E-09 2.6161E-09 1.5637E-09 9.5195E-10 0.17777563 0.190681 0.23221334 0.05969773 0.02558846 0.00031174 9.1563E-05 3.8491E-05 1.9664E-05 4.7763E-06 3.3485E-06 2.9586E-07 8.3595E-08 1.5513E-08 700 800 900 1000 -600



The RF system gives a large reactive contribution, second only to the space-charge term. Because of the relatively low figure of merit Q, there is also a substantial resistive contributions. In our analysis we have not included the impedance at the operating frequencies, since these are expected to be controlled and compensated with the low-level RF loops.

#### The Total Coupling Impedance of the NSNS Accumulator Ring

The total coupling impedance, that is the sum of all the contribution of the components presented above is given in Figure 1 versus the harmonic number n. The breakdown of the various contributions is shown in Tables 15 and 16. It is seen that at most the imaginary part is 230 ohm in the low-frequency range and about 150 ohm elsewhere. By far the largest contribution is the longitudinal space-charge, followed next by the RF cavity system. All other vacuum chamber components gave negligible imaginary contribution. On the other end, the resistive contribution at most is 3.5 ohm and always very small when compared to the imaginary term. In the large-frequency range, n > 10, the real part does not exceed 1 ohm. The kicker magnets are the major resistive source, followed, in the order, by the pipe steps, the damper system, and the resistivity of the wall with the RF cavity system.

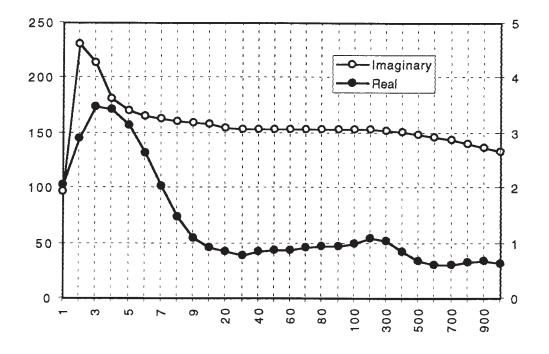


Figure 1. The Total Coupling Impedance in the NSNS Accumulator Ring

The analysis of which we have shown the results in this technical note does not include other special vacuum chamber items, specially those which are located in the injection region of the Accumulator Ring or other beam instrumentation devices. Table 15: Coupling Impedance of the various contribution. Real Part (ohm)

Real

Total		2.04201430	2.89/655/3	3.46621274	3.42476051	3.15202251	2.64059206	2.02990972	1.47210198	1.0820605	0 90987126	0 84070602	0 77172355	0.84787280	0.87404700		10410000.0	0.919/000	0.94/99114	0.94815159	U.90/94035	03016614	1.000400014	0.67002000	260505090	0.00614110	0.65754500	+1010100	0 66707470
RF Cavities	0 0000000000000000000000000000000000000					0.02558846	0.01357733	0.00813748	0.00528479 1	0.00363513	-	-					-			0.0400E-U0 C	-	R 3505E-08							
Vaccum Ports	0 86616 10	2.000 E-10	0.00436-09	2.00115-08	6.30/8E-08	1.232E-07	2.1288E-07	3.3805E-07	5.046E-07	7.1845E-07	9.8551E-07	7.8816E-06	2.6586E-05	6.2973E-05	0 00012288	0.00021200	0.00033633	0.00050126	0.0000000000000000000000000000000000000	0.0000752	0.0075602	0 02420912	0.0533133	001000000	0 14581208	0 20106822	0.25749809	30653060	
Steps	0 62641886	0.62641706	0 50541444	0.06040000	0.02040980	0.6264046	0.62639806	0.62639034	0.62638143	0.62637132	0.62636003	0.6261818	0.62588487	0.62546939	0.62493562	0 62428384	0.62351444	0 62262784	0.62162456	0.62050515	0.60309517	0.57515727	0 53820495	0 49416168	0 44519548	0 3935435	0.34134617	0 2005077	
Kicker	0.96753029	1 75097376	0 01060000	200001212	10211162.2	299/2920.2	1.52/3/903	0.94732133	0.43754353	0.10749521	1.5193E-32	3.0378E-32	4.5545E-32	6.0687E-32	7.5794E-32	9.0858F-32	1.0587F-31	1.2082F-31	1.35715-31	1 50515-31	2.9258E-31	3.6031E-30	5.222E-31	1.2036F-29	5.5779F-30	2.2351E-30	6.6239E-31	R DORFE-32	
Damper	0.11095551	0.21227678	0 20646		0.00400004	0.38445505	0.386201/1	0.36184107	0.31660948	0.2576075	0.19285211	0.1066706	0.0141331	0.0632297	0.06091484	0.02351443	0.04980194	0.05107936	0.02554442	0.04108165	0.01219357	0.00652138	0.00424504	0.0057506	0.00300037	0.00066587	0.00328405	0 00079594	
MHB	0.00340133	0.00680247	0.01020323	0.01360343		0.01/00/200	0.02040138	0.02379875	0.0271948	0.03058935	0.0339822	0.06777624	0.10119542	0.13405595	0.16617834	0.1973888	0.22752055	0.2564151	0.2839234	0.30990698	0.46463296	0.41629577	0.24410429	0.07867929	0.00359904	0.01436041	0.05237728	0 06699292	
Bellows	0	0	Ċ		> <	> <	о (	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	C	•
Resistive	0.15624336	0.11048042	0.09020646	0.07812057	0.06987966	0.06978907		6/160860.0	0.05523707	0.05207717	0.04940385	0.03492386	0.02850169	0.0246668	0.02204383	0.02010218	0.01858807	0.01736282	0.01634345	0.01547683	0.01063672	0.00828252	0.00671204	0.00551215	0.00453327	0.00371006	0.00301014	0.00241531	
Space Charge	0	0	0	C		> c	> <	2 (	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	
Free Space	0.00028996	0.00072421	0.00122525	0.00176123	0.00230921	0 00285038	0.00336867	0.00300000	0.00085039	0.00428411	0.00466073	0.00483389	0.00189032	0.00034958	3.2823E-05	1.6149E-06	4.2342E-08	5.9764E-10	4.5706E-12	1.9023E-14	1.3417E-52	1.281E-116	2.076E-206	0	0	0	0	0	
c	-	~	С	4	S	. (c	~ ~	- 0	0 0	ς Γ	0		30	40	50	60	70	80	06	100	200	300	400	500	600	700	800	006	

Table 16: Coupling Impedance of the various contribution. Imaginary Part (ohm)

s RF Cavities Total	3 -53 33269 96 5639625	79 9988076	61.9982133	29.3327115	17.6185757 1	11.8567153 1	8 8.55514662	5 6.47579103 1	B 5.07752812 158.985407		9 1.00516254 154.45638	0		6 0.15976164 153.630875		2 0.08129059 153.561296	2 0.06214306 153.494988	2 0.04901803 153.471166	0.03963106 1	1 0.00962819 153.033687	9 0.00408084 152.240003	3 0.00214797 150.880888	0.0012622 1	4 0.00078967 146.510982	1 0.00051285 143.687924	0.00034057	
Vaccum Ports	8.99356253		8.993531	8.9935038	8.99346855			8.99331585	8.99324928	8.99317488	8.99200009	8.99004176	8.98729939	8.98377226	8.97945947	8.97435992	8.96847232	8.961795	8.95432692	8.8356618	8.63517649	8.35024983	7.97972237	7.52562494	6.9945135	6.39807498	+ CY 0 C 3 C 3
Steps	-1.9679529	-1.9679473	-1.967938	-1.9679249	-1.9679081	-1.9678876	-1.9678633	-1.9678353	-1.9678035	-1.9677681	-1.9672082	-1.9662753	-1.96497	-1.9632931	-1.9612455	-1.9588284	-1.956043	-1.9528911	-1.9493744	-1.8946794	-1.8069099	-1.6908207	-1.5524547	-1.3986229	-1.2363534	-1.0723706	0 0106660
Kicker	-2.977752	-2.4090452	-1.6060225	-0.7444274	-1.241E-16	0.49627553	0.68826923	0.60222701	0.33083623	1.2407E-16	1.2403E-16	1.2397E-16	1.2389E-16	1.2378E-16	1.2366E-16	1.235E-16	1.2333E-16	1.2313E-16	1.2291E-16	1.1946E-16	3.3426E-16	1.0661E-16	4.3864E-16	2.5873E-16	1.4256E-16	6.7612E-17	2 044EE.17
Damper	-0.48178	-0.4366777	-0.3667539	-0.2791922	-0.1827727	-0.0868042	-1.747E-05	0.07043487	0.11974423	0.14588326	-0.0199328	0.02908987	0.05696027	0.00429716	0.0185588	0.03327659	-1.201E-17	0.00575771	0.0159905	-0.007616	0.0012311	-5.874E-19	0.00175415	-0.0026977	0.00160749	1.5312E-18	-0 00058
BPM	-0.6820807	-0.6820448	-0.681985	-0.6819014	-0.6817938	-0.6816623	-0.681507	-0.6813278	-0.6811247	-0.6808978	-0.6773215	-0.6713878	-0.6631366	-0.6526234	-0.6399188	-0.6251076	-0.6082885	-0.589573	-0.5690846	-0.3000918	-0:0311971	0.11014405	0.10387108	0.0238781	-0.0394549	-0.0462232	0.0152007
Bellows	-1.09784	-1.0978369	-1.0978317	-1.0978244	-1.097815	-1.0978035	-1.09779	-1.0977744	-1.0977567	-1.0977369	-1.0974245	-1.0969041	-1.096176	-1.0952405	-1.0940982	-1.0927498	-1.091196	-1.0894377	-1.0874758	-1.0569637	-1.0080007	-0.9432393	-0.8660506	-0.7802342	-0.6897107	-0.5982315	-0 5001337
Resistive	-0.1562434	-0.1104804	-0.0902065	-0.0781206	-0.0698726	-0.063784	-0.0590517	-0.0552371	-0.0520772	-0.0494038	-0.0349239	-0.0285017	-0.0246668	-0.0220438	-0.0201022	-0.0185881	-0.0173628	-0.0163434	-0.0154768	-0.0106367	-0.0082825	-0.006712	-0.0055121	-0.0045333	-0.0037101	-0.0030101	-0 0024153
Space Charge	148.266907	148.266846	148.266744	148.266603	148.26642	148.266197	148.265934	148.26563	148.265285	148.2649	148.25882	148.248686	148.2345	148.216263	148.193977	148.167643	148.137263	148.10284	148.064377	147.458385	146.453905	145.059118	143.285321	141.146777	138.660519	135.84613	132.725483
Free Space	-0.0001674	-0.0004181	-0.0007074	-0.0010168	-0.0013332	-0.0016457	-0.0019449	-0.002223	-0.0024734	-0.0026909	-0.0027908	-0.0010914	-0.0002018	-1.895E-05	-9.323E-07	-2.445E-08	-3.45E-10	-2.639E-12	-1.098E-14	-7.746E-53	-7.4E-117	-1.2E-206	0	0	0	0	c
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